

1 **Distinct degradation processes in ZnO varistors: reliability analysis and modeling**
2 **with accelerated AC tests**

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6 **Abstract:** In this study, we investigate different degradation mechanisms of ZnO
7 varistors. We propose a model showing how the V_v (defined as the DC varistor voltage
8 when 1mA DC current applied) changes by time for different stress levels. For this
9 purpose, accelerated degradation tests are applied for different AC current levels; then
10 voltage values are measured. Different from the common practice in the literature that
11 considers a degradation with only decreasing V_v values, we demonstrate either an
12 increasing or a decreasing trend in the V_v parameter. The tests show us a decreasing trend
13 in V_v for current levels above a certain threshold and an increasing trend for current levels
14 below this threshold. Considering both of these degradation mechanisms, we present a
15 mathematical degradation model. The proposed model exploits physics of the
16 degradations for a single grain boundary that is the core structure of ZnO varistors. To
17 validate the proposed model, we perform Monte Carlo simulations and the results are
18 compared with those obtained from accelerated AC tests. At the end as a summary of this
19 study, we introduce a conceptual accelerated AC test methodology to analyze the
20 reliability of a new ZnO varistor.

21 **Key words:** Reliability analysis and modelling, degradation, ZnO varistors, AC tests.

22
23 **1. Introduction**

1 Zinc oxide (ZnO) varistors have been widely used in electrical and electronics systems
2 against overvoltage surges thanks to their non-ohmic current-voltage characteristics and
3 excellent energy handling capabilities [1, 2, 3]. ZnO varistors are variable resistors used
4 for limiting or diverting transient AC line voltage; they are usually subjected to long-term
5 AC voltage and surge stresses which lead to degradation of them. Even varistors used
6 within their well-defined specifications might fail due to degradation; ageing
7 phenomenon makes the varistors degraded and even thermally broken down or destructed
8 [4]. Therefore degradation analysis of varistors is crucial.

9 Overwhelming majority of the studies in the literature have reported an increasing leakage
10 current and accordingly an increasing V_v parameter (defined as the DC varistor voltage
11 when 1mA DC current applied), manifesting the degradation [1, 5, 6, 7, 8]. On the other
12 hand, decreasing leakage current that results in an increase of V_v , is totally disregarded.
13 Only few studies have mentioned this without a detailed analysis [9, 10]. Here, the reason
14 is that as opposed to an increasing leakage current, a decreasing leakage current might
15 not hazardous for a varistor. However, we show and demonstrate that this kind of
16 degradation directly affect system reliability and safety [11]. This is illustrated in Figure
17 1. Even a slight increase in V_v leads to a dramatic change in maximum designed voltage
18 criteria for other components in the same block, especially for those in the power supply
19 block. This eventually causes a breakdown of the protected circuits and components. This
20 phenomena is demonstrated by testing over 100 electronic cards that were commercially
21 used and failed due to varistor degradations [11]. As a result, to our knowledge, this is
22 the first study on modelling and characterization of ZnO varistors regarding both
23 degradation processes.

1 We perform accelerated AC voltage/current tests to analyze varistor degradations. In
2 general, accelerated tests are conducted by applying stresses such as temperature,
3 vibration, humidity, and voltage, beyond their normal/expected levels in field [12, 13].
4 We use a single stress variable of voltage since it is by far the most dominant stress for
5 varistors. Indeed, there are two major reliability/degradation test methodologies for
6 varistors that are pulse tests (8/20 μ s and 2ms tests) and accelerated AC tests. We use
7 accelerated AC tests since it is cost efficient both in terms of test time and test equipment.
8 In the literature there are studies using AC test with showing that that the V_v parameter
9 of varistors which is one of the key parameters related to measuring reliability of a
10 varistor, can be changed during the tests [7, 11, 14]. In this study, we make reasoning,
11 justification, and modelling of these V_v changes. By applying accelerated AC signals, we
12 see either an increasing or a decreasing trend in V_v that depends on the current levels
13 passing through the varistor. The tests show a continuous and fast decreasing trend in V_v
14 for current levels above a certain threshold and a slow increasing trend for current levels
15 below the threshold. Although we apply accelerated AC tests to a particular metal-oxide
16 varistor -- detailed information such as composition, microstructure, and phases of this
17 particular ZnO varistor is available in the data sheet [15], our results are generally
18 applicable to variety of similar metal-oxide varistors.

19 We investigate physical bases of the degradation processes observed in the AC tests.
20 Physically, a ZnO varistor has a structure containing numerous number of zinc oxide
21 grains of different shape and size with other metal oxide additives [15]. Several studies
22 demonstrate that electrical stresses on varistors cause deformation of grain boundary
23 potential barriers which are highly resistive [5, 16, 17]. It is reported that mild degradation
24 after applying an electric field stress leads to a decrease in the effective doping

1 concentration of grains and accordingly an increase in the resistance of grain boundaries
2 [6]. On the other hand, applying a strong electrical field might lead to an increase in the
3 doping concentration resulting in a decrease in the resistance of grain boundaries [18].
4 Motivated by this, in our model, we assume that every single ZnO-ZnO grain boundary
5 can be considered as a resistor. We formulate the resistor as a function of time and applied
6 stress that determines relative effects of the degradation processes. Additionally, we need
7 to determine how to combine the resistors by considering grain shapes and amounts. In
8 this regard, several studies have been proposed aiming on varistor microstructures by
9 considering grain shape and topology as well as type and distribution of defects [19, 20,
10 21]. In a similar way, we present a simple yet accurate microstructure that solely consists
11 of grain boundary resistors. We perform a Monte Carlo method to validate our model;
12 outcomes of the simulations match well with the results obtained from the accelerated
13 AC degradation tests.

14 The outline of paper is as follows. In Section 2, the proposed accelerated AC test
15 methodology used to distinguish degradation processes is presented. In section 3, the
16 physical bases of the processes are presented and used to form a varistor degradation
17 model. Simulation results are also given to validate the proposed model. In the section 4,
18 a reliability test methodology is presented to facilitate and summarize the steps proposed
19 in this study. Section 5 reports the conclusion of this work.

20

21 **2. Degradation processes by accelerated AC tests**

22 In order to investigate varistor degradation mechanisms, we perform accelerated AC tests
23 which are cost efficient both in terms of test time and test equipment. The tests show
24 different degradation processes including stable, decreasing, and increasing trends in V_v

1 depending on the applied AC signal levels. Different accelerated AC signal levels are
2 applied on varistor samples having a diameter of 12mm and a height of 15mm as shown
3 in Figure 2. We number varistors starting from Var1 to Var11 which samples are of the
4 same material, manufacture, and characteristics [15].

5 For a specific AC current passing through a varistor called as I_{VAR} (RMS value), voltage
6 values of the varistor called as V_{VAR} (RMS value) are periodically measured. Our three-
7 step test procedure is summarized as follows:

8 1) All samples are shorted to ground for 24 hours in order eliminate any previous
9 capacitive/inductive loads.

10 2) I_{VAR} is determined and continuously applied. Then V_{VAR} 's are periodically
11 measured for a 300 minute time period. If the varistor burns then the tests are
12 stopped at the time of the break down (failure).

13 3) Step-2 is repeated for different I_{VAR} values to analyze different degradation
14 mechanisms and finally to find the I_{VAR} threshold value.

15 For the second step, we need to determine which I_{VAR} values are used. Indeed, the
16 suggested I_{VAR} value according to IEC standard [IEC-60099-4] is 1mA [22]. However,
17 we show that applying different I_{VAR} values results in different degradation mechanisms.
18 This is illustrated in Figure 3 with using 6 different I_{VAR} values 5mA, 3mA, 2mA, 1.5mA,
19 1mA, and 0.7mA. Note that for a certain I_{VAR} value, multiple samples are generally used
20 to take into account probable process variations. Analyzing the results in Figure 3, we
21 certainly see two different degradation mechanisms corresponding to increasing and
22 decreasing V_{VAR} values over time. While a decrease in V_{VAR} results in an increase in
23 leakage current and a decrease in V_V , increasing V_{VAR} makes the leakage current decrease
24 and V_V increase. Note that a varistor voltage V_V represents a varistor voltage when 1mA

1 DC current is conducted. Thus there is a positive correlation between V_{VAR} and V_V if a
2 considerable amount of varistor current is conducted in the range of mA's. In the
3 following section we separately analyze the cases in Figure 3 to find the threshold value
4 of I_{VAR} that distinguishes between the degradation mechanisms.

5

6 **Finding the threshold:**

7 In order to find the threshold value, we first tried $I_{VAR} = 10\text{mA}$. Here, the varistor voltage
8 plunged dramatically and after seconds varistor started to burn up. We decreased this
9 value to 5mA but still there was a dramatic decline in the varistor voltage. Then we
10 performed $I_{VAR} = 2\text{mA}$ tests for two samples as shown in Figure 4(a). Here, we do not
11 have not full data covering 300 minutes since the varistors started to burn up. We
12 continued reducing I_{VAR} until V_{VAR} values stabilized by time. This was achieved when
13 $I_{VAR} = 1.5\text{mA}$. Indeed as shown in Figure 4(b), when $I_{VAR} = 1.5\text{mA}$, V_{VAR} shows a slight
14 decrease. It means that the threshold value is slightly under 1.5mA. We repeated the test
15 for $I_{VAR} = 1\text{mA}$ and $I_{VAR} = 0.7\text{mA}$ as shown in Figure 5 (a) and (b), respectively.

16 In conclusion, for current levels above 1.5 mA which is found the threshold value for this
17 family of varistor samples, V_{VAR} drops dramatically until it burns. We classify this as a
18 hard degradation mechanism. On the other hand, for current levels below the threshold
19 value (1.5 mA), we see a relatively slow increasing trend in V_{VAR} . We classify this as a
20 mild degradation mechanism.

21

22 **3. Physical bases of degradation processes and mathematical modeling**

23 In this section, we aim to develop a mathematical reliability model for varistor
24 degradations. For this purpose we first investigate physical bases of the degradation

1 processes observed in the AC tests. It is a general conception that electrical stresses on
2 varistors cause deformation of grain boundary potential barriers [5, 16, 17]. Figure 6
3 shows a conceptual microstructure of a ZnO varistor where grains and boundaries are
4 represented with white and grey regions, respectively. If boundaries are overcome then a
5 current is conducted through paths of grains that is illustrated with the arrows in the
6 figure. Note that while ZnO grains are conductive, intergranular boundaries are highly
7 resistive. Since ZnO grains and boundaries are spread in almost a regular fashion, a single
8 grain boundary between two grains can be used as a core structure for varistor models.
9 This is indeed called a single grain boundary model [23, 24].

10 A ZnO-ZnO grain boundary can be modeled with a Schottky barrier as shown in Figure
11 7. By depletion of carriers from surrounding grains, a double Schottky barrier is forming.
12 Long-term stresses cause a degradation along with a change in the barrier height and
13 characteristics due to migration of ionized donors in electric field and redistribution of
14 them in the near-surface region of grains. This results in an increase in leakage current,
15 relevantly a decrease in V_v [25].

16 Other possible mechanisms of deformation of grain boundary barriers explain different
17 processes of degradation in ZnO varistors. It is reported that there are two types of
18 electron traps of importance for ZnO varistors: traps located at ZnO-ZnO grain
19 boundaries known as interface traps, and traps located within the bulk of ZnO grains
20 known as bulk traps. Interface trapping of electrons is generally considered as the
21 mechanism giving rise to double Schottky barriers at grain boundaries [26].

22 Furthermore, it is reported that mild degradation after applying an electric field stress
23 leads to a decrease in the effective doping concentration at the grains and accordingly an
24 increase in the resistance of grain boundaries [6]. This increase clearly explains an

1 increase in V_v . On the other hand, applying a strong electrical field might lead to an
2 increase in the doping concentration resulting in a decrease in the resistance of grain
3 boundaries [18]. Subsequently, it results in a decrease in V_v .

4 Motivated by the mentioned studies on physics of degradation demonstrating the change
5 in grain boundary resistance values for different stress types, we model every single grain
6 boundary barrier as a resistor. The resistor value can change according to applied stress
7 in either increasing or decreasing trend. Thus, in this study, we contribute to a non-linear
8 cross boundary resistor as a function of time and stress. This is illustrated in Figure 8.

9 The proposed model formula for a single resistor (grain boundary) is shown below:

$$10 \quad R_t(t, s) = R_{in}(1 + \alpha R_i(t)R_i(s) + \beta R_d(t)R_d(s)) . \quad (1)$$

11 Degradation depends on both stress (s) and time (t). The initial value of the resistor,
12 degradation free value, is represented by R_{in} that is obtained using initial V_{VAR} and I_{VAR}
13 values. Degradations mechanisms causing increasing V_v and decreasing V_v in time
14 domain are represented by $R_i(t)$ and $R_d(t)$ functions, respectively. We select $R_i(t) = \exp(-$
15 $nt^{1/m})$ used for AC signal stresses. We select $R_d(t) = 1/\sqrt{t}$ that is derived using the leakage
16 current formula given in [16]. Although there is a common consensus in the literature for
17 decreasing V_v behavior in time domain shown by $R_d(t)$, increasing V_v behavior and
18 corresponding physical models lack precision. Therefore, in determination of $R_i(t)$, our
19 main source is the test data obtained from our experiments. The functions $R_i(s)$ and $R_d(s)$
20 represent the effects of applied stresses on degradation. They could be in the form of the
21 following exponential/power functions [27].

- 22 • Arrhenius $R(s) = e^{c/s}$
- 23 • Eyring $R(s) = s^{-1} e^{c/s}$
- 24 • Inverse power law $R(s) = s^n$

1 Note that since the above equations are used for a single stress (s), they can be applicable
2 for our case using $s = I_{VAR}$. In our accelerated tests we use a single stress of current I_{VAR} .
3 According to the tests described in Section 2, the threshold current (I_{VAR-th}) is close to 1.5
4 mA that determines the effects and dominances of $R_i(s)$ and $R_d(s)$ for different I_{VAR}
5 levels. Regarding this, we select $R_i(s) = (1 - e^{-cs})$ and $R_d(s) = e^{c(s - I_{VAR-th})}$ where $s = I_{VAR}$. The
6 coefficients in the degradation formula α , β , n , m , and c are empirically calculated using
7 the test data. As a result: $\alpha = 0.003$, $\beta = 0.001$, $n = 0.1$, $m = 0.7$, and $c = 4000$.

8

9 **Varistor microstructure formation and simulation results:**

10 Varistor ceramics are compound from ZnO grains of different shape, size, and orientation.
11 Since ZnO grains have predictable sizes, the total number of the grains in a varistor can
12 be found with using the varistor size [15]. Suppose that each grain diameter is in the range
13 between 10 μm and 100 μm and a ZnO varistor has a size of nearly 5mm \times 10mm in two
14 dimensions [15] (the test samples) that results in 5 Thousand to 0.5 Million grains. Using
15 these relatively high number of grains or grain boundaries and corresponding non-linear
16 resistors in simulations is certainly unpractical, so we need a microstructure. Figure 9
17 shows a simplified varistor microstructure based on equal cubic grains or square grains
18 on a mesh which represents all grains and boundaries [19, 20]. We exploit this structure
19 by using resistors only for grain boundaries.

20 Considering the tested varistor dimensions, we use X and $3X$ number of vertical and
21 horizontal squares, respectively. Here, increasing the value of X results in better accuracy
22 at the cost of worse runtime and complexity for the simulation. Therefore we need to
23 determine the minimum value of X for which we achieve relatively high accuracy. For
24 this purpose we start with $X=1$, and increase X one by one. It is apparent that for $X=1$,

1 there is no vertical grain boundary. For $X=2$ and $X=3$, corresponding microstructures are
2 shown in Figure 10. For each case we perform a Monte Carlo analysis using multiple
3 samples. This is illustrated in Figure 11. For all four graphs in the figure, we use the test
4 data obtained using $I_{VAR} = 2\text{mA}$. According to Figure 11, as expected increasing X or the
5 number of resistors improves the simulation accuracy that depends on the fluctuations
6 between samples as well as the curve fitness to the test data. Analyzing the results in the
7 figure, we conclude that the cases $X=3$ (26 resistors) and $X=4$ (58 resistors) have almost
8 similar performances. Trying larger X values also give very similar results. Therefore we
9 decide to use $X=3$ with 58 resistors in further simulations.

10 Figure 12 shows the results comparing the test data obtained using I_{VAR} 's of 0.7mA, 1mA,
11 1.5mA, and 2mA and the related curves obtained using the proposed microstructure
12 having 58 resistors. Note that since the varistor started to burn up after 130 minutes for
13 2mA current stress, there is limited test data for this stress level. Each of the 58 resistors
14 is treated independently using the formula in (1). The results clearly proves the accuracy
15 of the proposed model; there is almost a perfect match between the curves and the real
16 test points.

17

18 **4. Reliability test methodology for a new varistor**

19 A variety of endurance and environmental tests are conducted to assure the reliability of
20 ZnO varistors. These tests are derived from the extremes of expected application
21 conditions, with test conditions intensified to obtain authoritative results within a
22 reasonable period [14]. The most commonly used test methodologies include “surge
23 current derating (8/20 μs)”, “surge current derating (2 ms)”, “fast temperature cycling”,
24 and “vibration” tests. Usually these methodologies are costly in terms of test time and test

1 equipment. Additionally, only the FAIL/PASS criteria is considered for most of these
2 methodologies, so they lack of analyzing reliability performance of varistors in time
3 domain. They also neglect different varistor degradation mechanisms explained
4 thoroughly in the previous sections. Considering these drawbacks, we present a simple
5 yet efficient test methodology to assess the reliability of a ZnO varistor based on
6 accelerated AC degradation tests. Figure 13 shows a flowchart of the proposed
7 methodology. According to the flowchart, it is possible to analyze a varistor reliability
8 based on accelerated degradation AC tests with using the proposed degradation model.
9 First, accelerated AC tests are applied to find the threshold value that distinguishes
10 between the degradation mechanisms. As examined in Section 2, the threshold value of
11 ZnO varistors should be close to a 1 mA current. Second, both heavy and moderate
12 degradation tests are performed to specify the trend of the mechanisms. Third, the
13 proposed degradation model is used. The test data is fitted to the model by calculating
14 empirical coefficients from the experiments. By considering the obtained degradation
15 formula and the tested varistor usage in field, different critical degradation values can be
16 chosen by designers. Finally, varistor reliability can be estimated.
17 Note that the proposed methodology is conceptually given without certain procedures and
18 steps having specific numbers for different varistor families. These details are out of
19 scope of this study and can be considered as a future work.

20

21 **5. Conclusion**

22 In this paper, we study degradation processes for ZnO varistors. For this purpose,
23 accelerated AC degradation tests are applied aiming on measuring varistor voltage values
24 in time domain. As opposed to the common practice in the literature that considers a

1 degradation with only decreasing V_v values, the tests show either an increasing or a
2 decreasing trend in the V_v parameter. To justify the observed degradation processes,
3 physical bases of the degradations are investigated. For this purpose, a single grain
4 boundary (Schottky barrier) is modeled as a non-linear resistor; its characteristics in time
5 domain changes with applied stress levels. Then a microstructure is formed using these
6 boundary resistors.

7 We perform a Monte Carlo method to validate the proposed varistor degradation model;
8 outcomes of the simulations match well with the results obtained from the accelerated
9 AC degradation tests. As a summary of this study, we introduce a conceptual accelerated
10 AC test methodology to make the reliability analysis of a new ZnO varistor.

11

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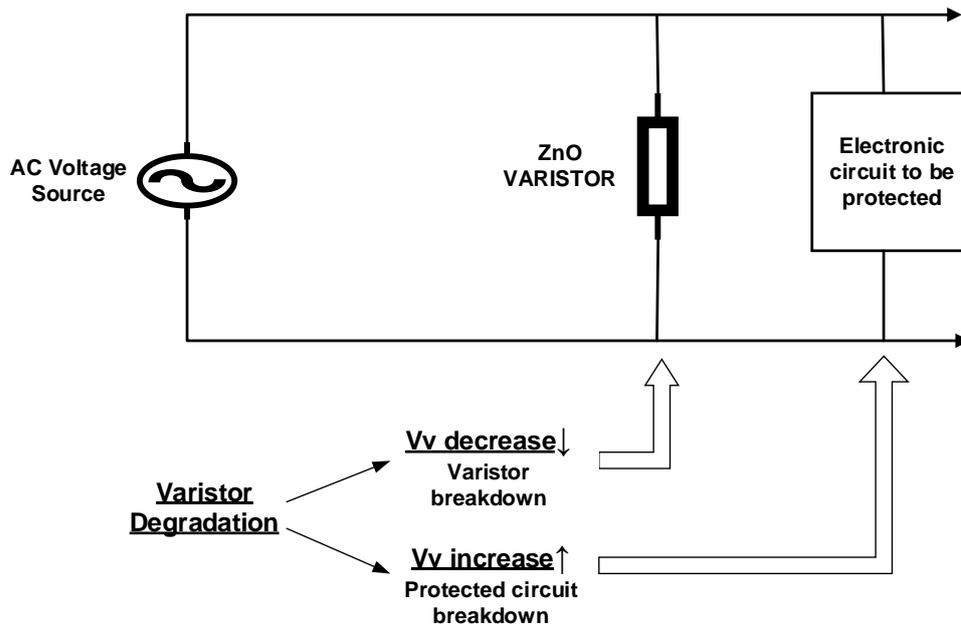
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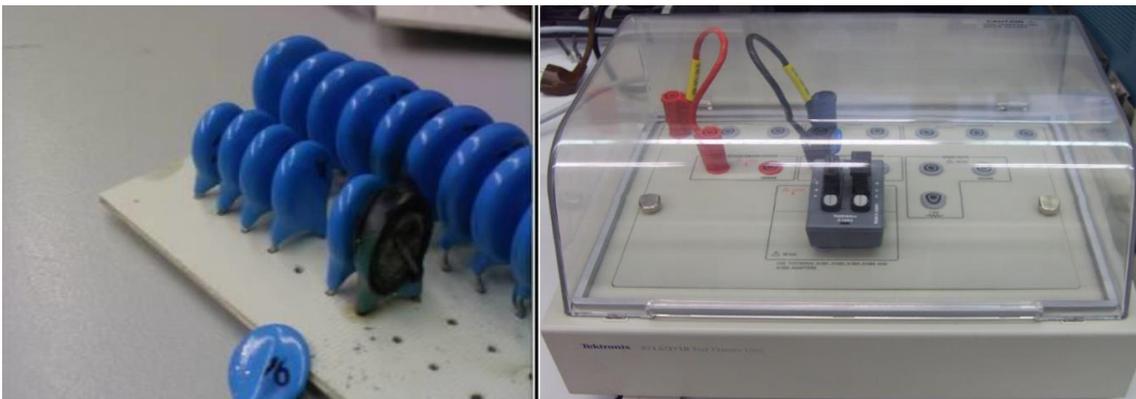


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2 **Figure 1.** Breakdown of the system in case of different degradation processes of ZnO

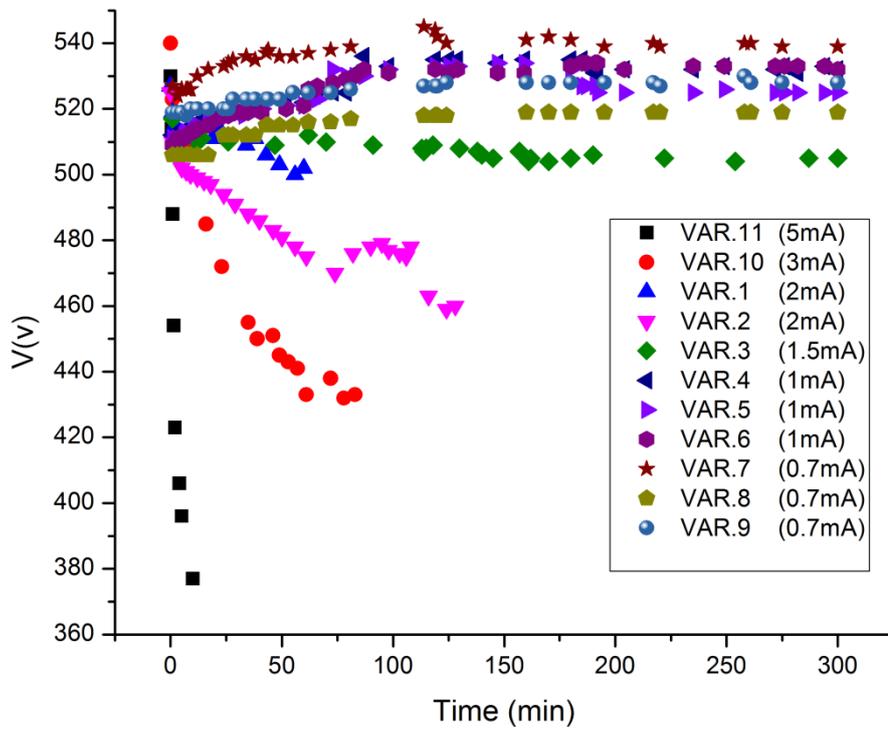
3 varistor[11].

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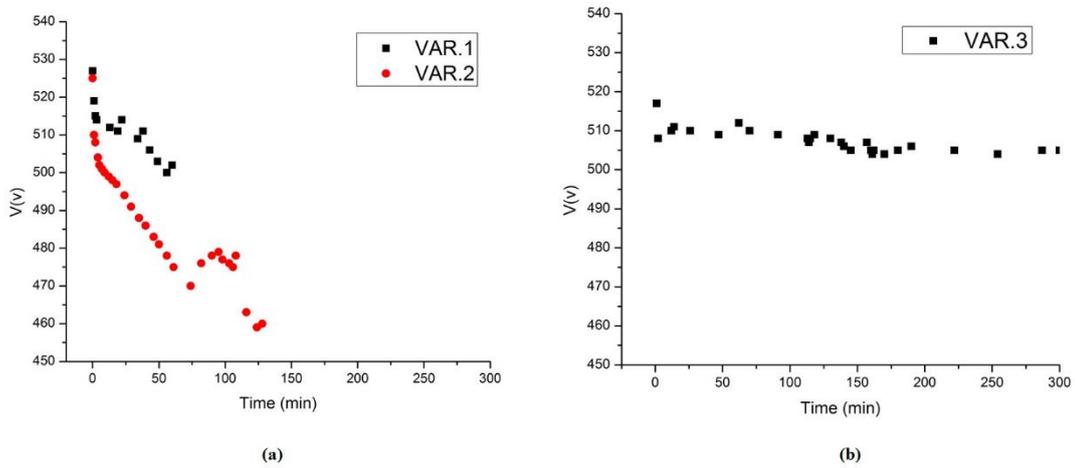
6 **Figure 2.** Varistor samples used in the accelerated AC tests.



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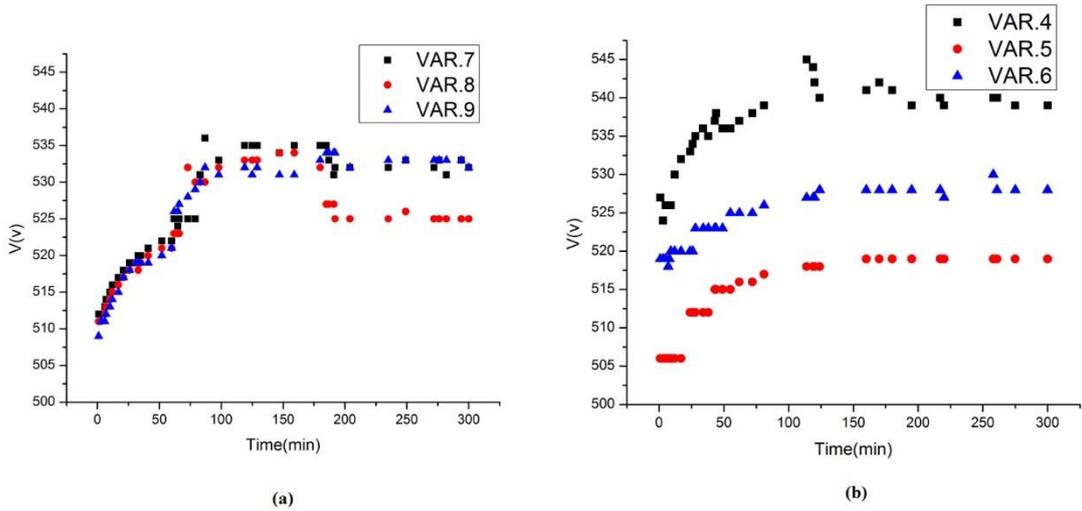
2 **Figure 3.** Accelerated degradation tests results: V_{VAR} values in time domain for different
 3 I_{VAR} values.

4



5

6 **Figure 4.** Accelerated degradation tests results: V_{VAR} values in time domain for (a) I_{VAR}
 7 $= 2\text{mA}$ and (b) values $I_{VAR} = 1.5\text{mA}$.

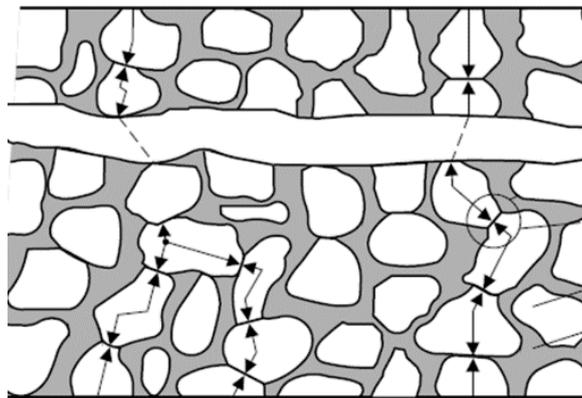


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2 **Figure 5.** Accelerated degradation tests results: V_{VAR} values in time domain for (a) I_{VAR}

3 $= 1\text{mA}$ and (b) values $I_{VAR} = 0.7\text{mA}$.

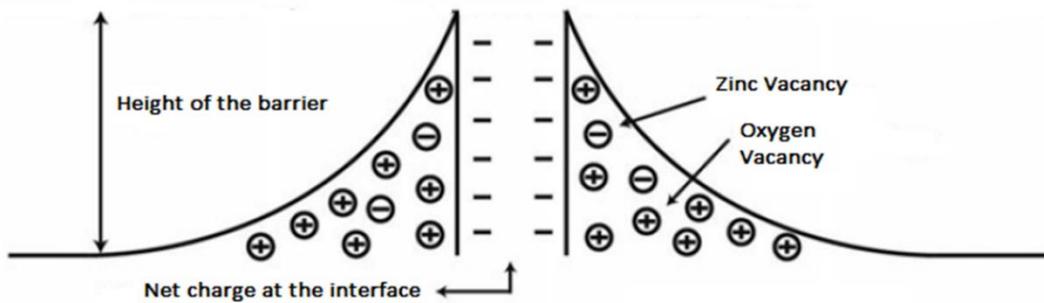
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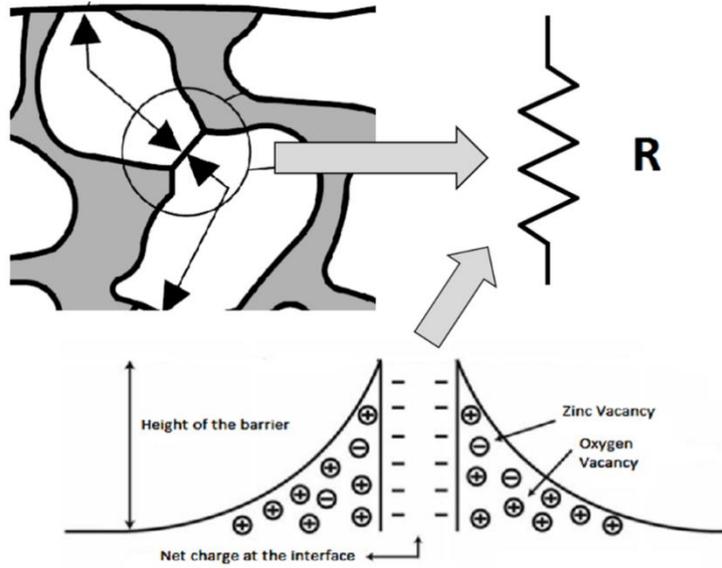
6 **Figure 6.** Microstructure of a varistor element [15].

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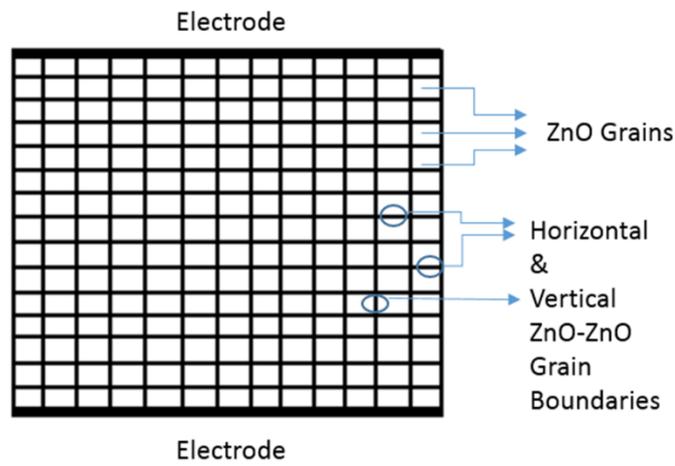
9 **Figure 7.** A simple double Schottky barrier model of a ZnO-ZnO grain boundary.



1

2 **Figure 8.** Modeling of a single grain boundary as a resistor.

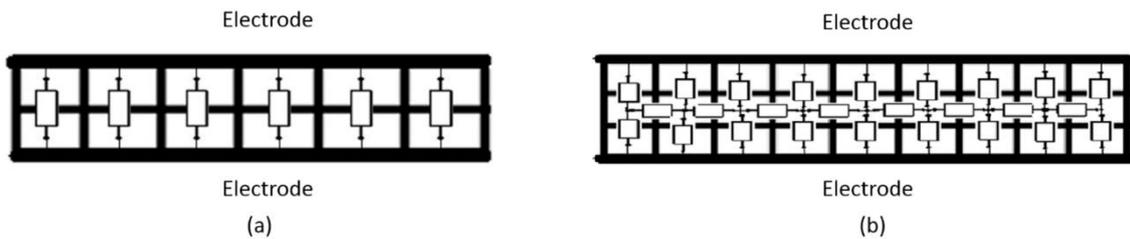
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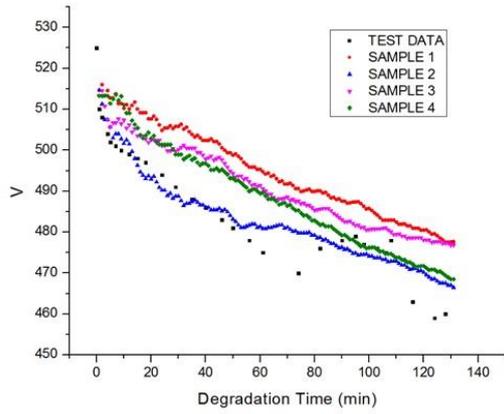
5 **Figure 9.** Simplified grain model of a varistor.

6

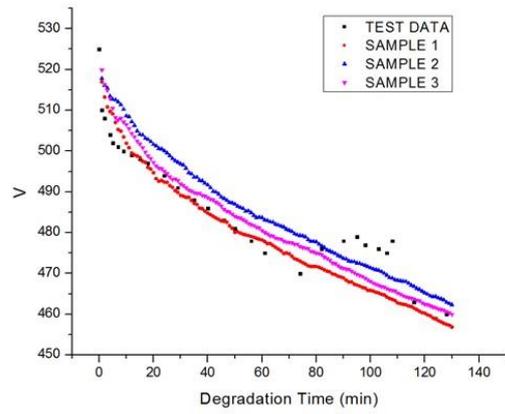


7

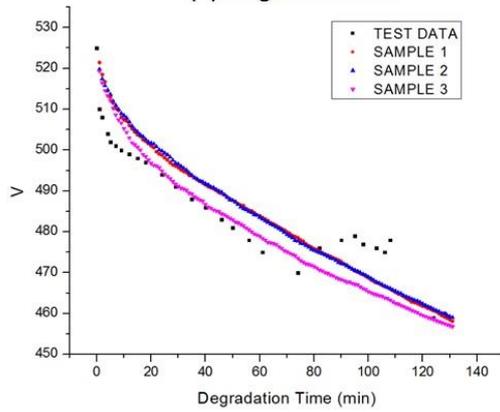
8 **Figure 10.** Proposed microstructures (a) $X=2$, 6 resistors and (b) $X=3$, 26 resistors.



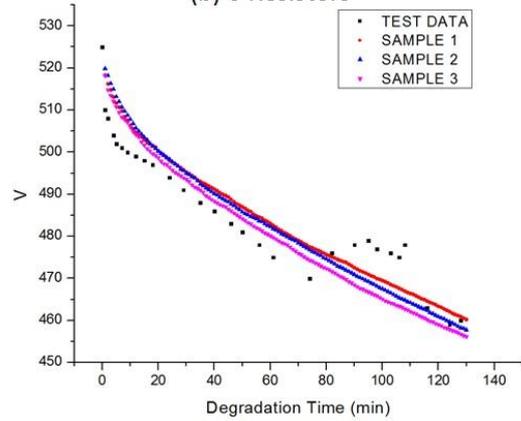
(a) Single Resistor



(b) 6 Resistors



(c) 26 Resistors

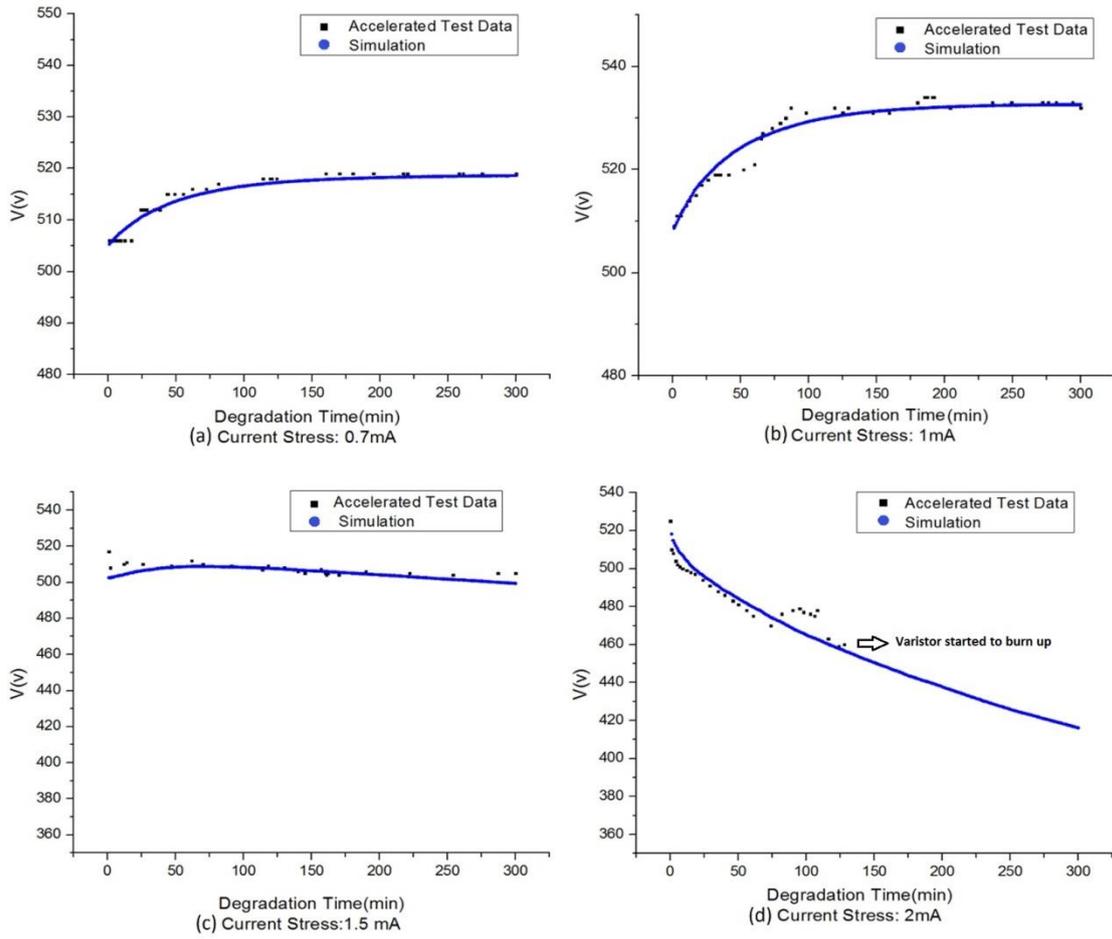


(d) 58 Resistors

1

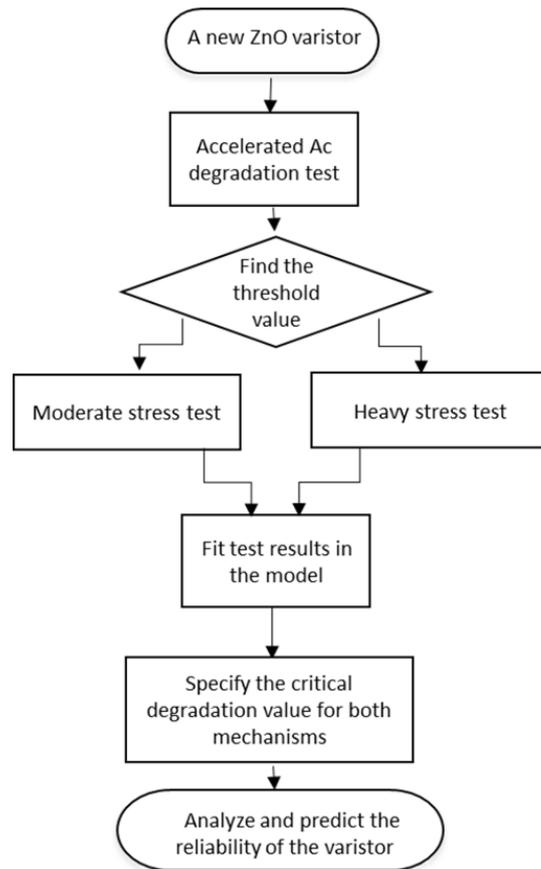
2 **Figure 11.** Simulation results in comparison with the AC test data for (a) single resistor,

3 (b) $X=2$, 6 resistors, (c) $X=3$, 26 resistors, and (d) $X=4$, 58 resistors.



1

2 **Figure 12.** Accelerated degradation test data in comparison with the simulation results
 3 for (a) $I_{VAR} = 0.7\text{mA}$, (b) $I_{VAR} = 1\text{mA}$, (c) $I_{VAR} = 1.5\text{mA}$, and (d) $I_{VAR} = 2\text{mA}$ current
 4 stresses.



1

2

Figure 13. Reliability test methodology flowchart for ZnO varistors.